



Terminating a differential-input signal

A PREVIOUS COLUMN explains that terminating transmission lines at the driving and receiving ends minimizes signal reflections (Reference 1). You must calculate termination-resistor values for single-ended and

fully differential circuits. The calculations for single-ended circuits are simple; the noninverting-circuit configuration separates the termination and gain-setting resistors. The calculations in fully differential circuits are complicated, because you can't separate the termination and gain-setting resistors (Figure 1).

Assuming that you're dealing with an ideal op amp simplifies the calculations; doing so makes the amplifier gain infinite with no frequency degradation. Under these assumptions, $V_N = V_P$, causing a virtual short across the op-amp inputs. Thus, R_1 is effectively in series with R_3 , and the series combination is in parallel with R_T . The equation for the terminating-resistor value follows:

$$R_T = \frac{R_S(R_1 + R_3)}{(R_1 + R_3 - R_S)}$$

where R_S is the driver/source termination resistor.

You must account for the source output impedance, R_S , and the cable-termination resistance, R_T , in the gain equation, because they

are parts of gain-setting resistors R_1 and R_3 (R_G in the general case). Begin the gain-resistor calculations by using Thevenin's theorem to obtain a series-equivalent circuit to replace R_T and R_S . Thevenin-equivalent voltage and resistance equations for a series model follow:

$$V_{TH} = V_S \frac{R_T}{R_T + R_S}$$

$$R_{TH} = \frac{R_S R_T}{R_S + R_T}$$

Substituting these three equations into the ideal gain equation, $V_{OUT} = V_{IN} (R_F/R_G)$, yields the gain equation for a circuit with significant source/termination impedance. This equation applies to fully differential inputs and outputs; hence, $R_G = R_1 = R_3$, and $R_F = R_2 = R_4$.

$$\frac{V_{OUT}}{V_{IN}} = \frac{R_F}{R_G + \frac{R_S R_T}{2(R_S + R_T)}} \frac{R_T}{R_T + R_S}$$

A design example calculates resistor values for a common circuit with a 50Ω differential-balance source and an overall differential gain of one (Reference 2).

First, choose a value for R_G , use that value to calculate R_F , and then use the values for R_T and R_G to calculate the feedback-resistor value. An appropriate feedback-resistor value for an op amp in this frequency range is 500Ω . R_T approximately equals R_S , so the gain through the termination resistors is about 0.5. The op-amp gain must be two to achieve an overall gain of one, so begin the calculations with $R_G = (R_F/2) = 249\Omega$ (closest 1% standard value). Now, use the first equation to calculate $R_T = 55.6\Omega$ and select R_T as the closest 1% standard value, 56.2Ω . An algebraically manipulated equation follows, which solves for $R_F = 495.5\Omega$, and the closest 1% standard value is 499Ω .

$$R_T = \left(R_G + \frac{R_S R_T}{2(R_S + R_T)} \right) \left(\frac{R_S + R_T}{R_T} \right)$$

Figure 2 shows the complete circuit with the calculated resistor values. □

REFERENCES

1. Mancini, Ron, "Fully differential amplifiers and transmission lines," *EDN*, Aug 7, 2003, pg 20.
2. Karki, James, *Fully Differential Amplifiers*, SLOA054D, Texas Instruments, January 2002.

Ron Mancini is staff scientist at Texas Instruments. You can reach him at 1-352-569-9401, rmancini@ti.com.

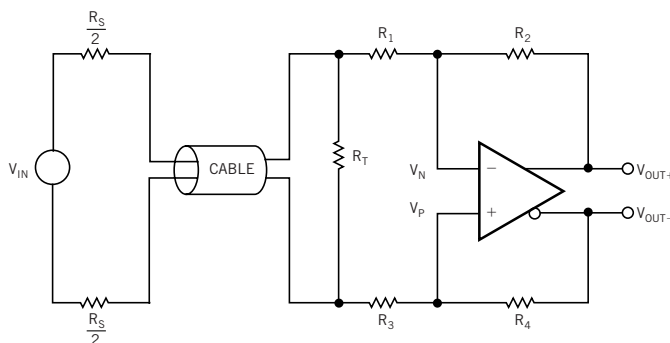


Figure 1 Fully differential termination is more complicated than the single-ended scheme, because you must take the gain-setting resistors into account when doing the calculations.

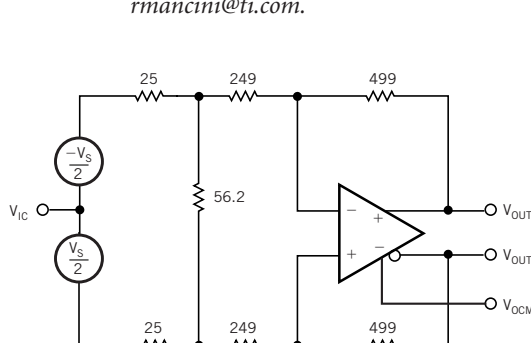


Figure 2 The calculated resistor values for the completed circuit differ from the intuitive values because R_T and R_S affect the calculations.