



# Low-voltage reference uses the $\Delta V_{BE}$ circuit

SOME OF MY RECENT COLUMNS discuss the functions and advantages of zener-diode references (references 1, 2, and 3). Although zener diodes are stable and precise voltage references, they require high bias voltage (usually 8V minimum). The bias-voltage requirement precludes most zener-diode references from circuits with

supply voltages of 5V or less. Low-voltage zener diodes are available, but their temperature drift renders them useless in precision applications. All semiconductor structures yield basic circuits that are temperature-sensitive; thus, any semiconductor reference must

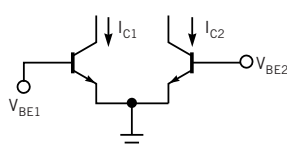
follow (Figure 1):

$$\Delta V_{BE} = V_{BE1} - V_{BE2} = \frac{kT}{q} \ln(J_{E1}) - \frac{kT}{q} \ln(J_{E2}) = \frac{kT}{q} \ln\left(\frac{J_{E1}}{J_{E2}}\right), \quad (1)$$

## $\Delta V_{BE}$ CIRCUITS HAVE THE SAME TEMPERATURE-DRIFT PROBLEM AND SOLUTION [THAT ZENER-DIODE REFERENCES HAVE], BUT BECAUSE THEY DO NOT DEPEND ON A ZENER JUNCTION, THEY FUNCTION AT LOWER VOLTAGE.

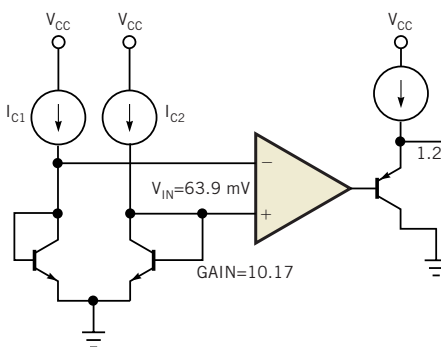
have some form of temperature compensation built in. Zener-diode references use an internal series diode for temperature compensation. Base-emitter-offset-voltage circuits have the same temperature-drift problem and solution, but because they do not depend on a zener junction, they function at lower voltage.

The difference equations for two transistors connected as diodes



**Figure 1**  $\Delta V_{BE}$  causes a current ratio in the transistors.

where  $\Delta V_{BE}$  is the base-emitter offset voltage,  $k$  is Boltzmann's constant,  $q$  is the electron charge,  $T$  is the absolute temperature,  $\ln$  is the natural log, and  $J_E$  is the emitter



**Figure 2** You amplify the  $\Delta V_{BE}$  and add it to a pnp transistor for  $V_{BE}$  to form a bandgap voltage reference.

current density. When the units of current density are equal, they cancel each other out, and Equation 1 reduces to Equation 2, which expresses the temperature coefficient as a function of the collector currents.

$$\Delta V_{BE} = V_{BE1} - V_{BE2} = \frac{kT}{q} \ln\left(\frac{I_{C1}}{I_{C2}}\right). \quad (2)$$

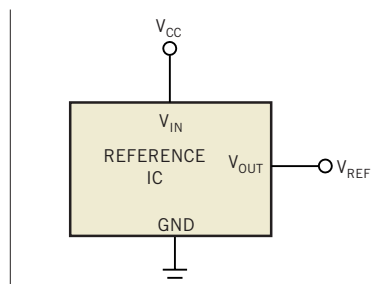
You obtain the temperature coefficient by differentiating Equation 2.

$$TC = \left(\frac{d\Delta V_{BE}}{dT}\right) = 86.3 \frac{\mu V}{^\circ C} \ln\left(\frac{I_{C1}}{I_{C2}}\right). \quad (3)$$

You use a diode-base-emitter junction to cancel out the drift that Equation 3 calculates, so the total drift must equal a diode drift of 2.18 mV/°C; the corresponding offset is 650 mV. You can't achieve the area ratio you need to meet these criteria through dissimilar junction areas because this ratio is too large; thus, use a smaller ratio, such as 12-to-1, and the resulting offset voltage is 63.9 mV. An amplifier with a gain of 10.17 yields an out-

put voltage of 650 mV (Figure 2). Multiplying the  $214\text{-}\mu\text{V}/^\circ\text{C}$  result of Equation 3 by the 10.17 amplifier gain yields a drift of  $2.18\text{ mV}/^\circ\text{C}$ .

The complete bandgap reference uses two transistors for the diodes; these transistors feed current in a 12-to-1 ratio to generate the basic offset voltage and temperature coefficient. An amplifier that



**Figure 3**

This voltage-regulator schematic resembles the schematic for a  $\Delta V_{BE}$  circuit.

increases the offset voltage to 650 mV follows this stage. The output voltage drives the base of a pnp transistor, which has a current source at its emitter. The output voltage is the sum of the pnp base-emitter voltage and the multiplied differential-diode voltage: 1.25V. The final result is that the pnp base-emitter temperature coefficient cancels out the differential-diode temperature coefficient, yielding a stable, low-drift voltage reference. The output stage buffers the voltage reference and provides gain or attenuation to generate different reference voltages. This voltage reference functions in a low-voltage environment with temperature stability and low drift. In schematics, base-emitter-offset-voltage references often look similar to voltage regulators, such as the one in Figure 3, but don't confuse them with voltage regulators,

because the references have limited current capacity. □

#### REFERENCES

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3. Mancini, Ron, "The ultimate zener-diode reference," *EDN*, Sept 30, 2004, pg 26.

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